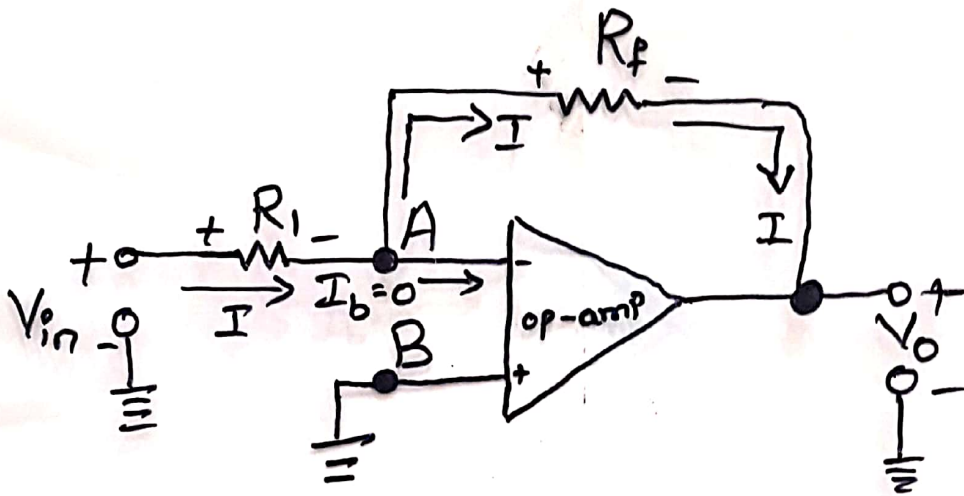


# Ideal Inverting Amplifier :-

(1)



## Derivation of closed loop gain :-

→ As node B is grounded, node A is also ground potential, from the concept of virtual ground, so  $V_A = 0$

$$\therefore I = \frac{V_{in} - V_A}{R_1}$$

$$I = \frac{V_{in}}{R_1} \quad \text{--- (1)}$$

→ Now from the output side, considering the direction of current (I) we can write,

$$I = \frac{V_A - V_O}{R_f}$$

$$I = \frac{-V_O}{R_f} \quad \text{--- (2)}$$

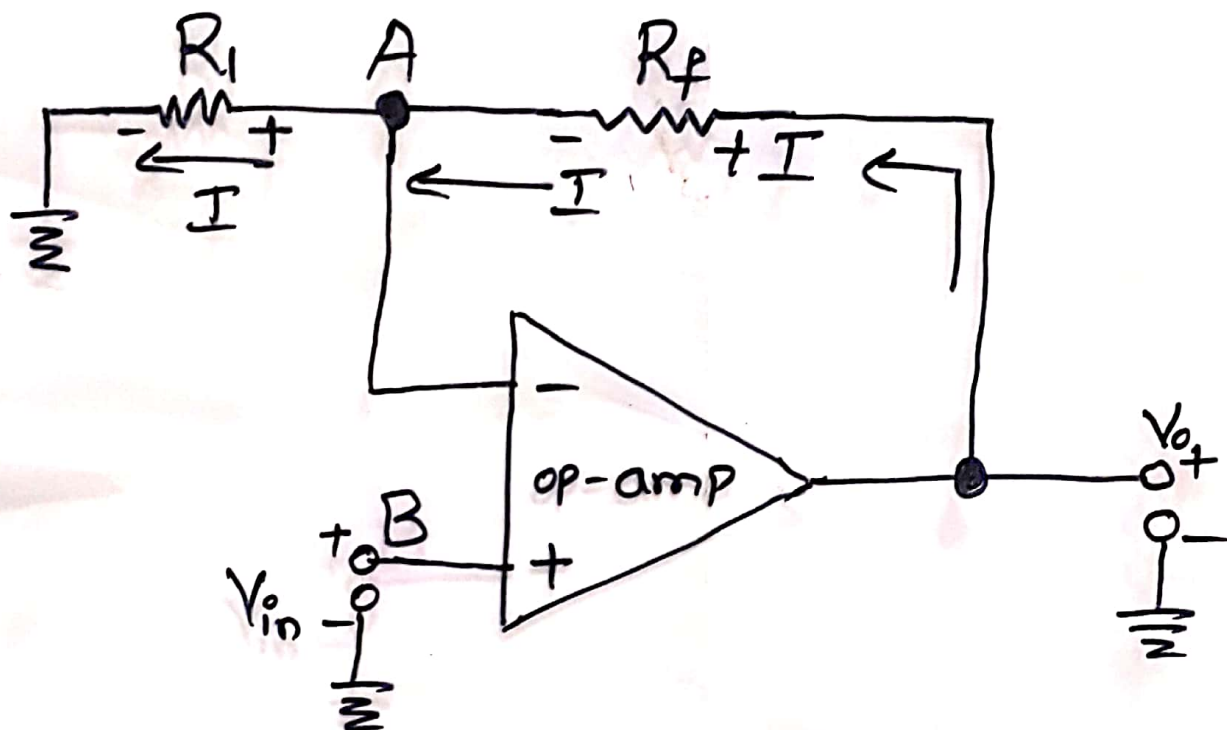
Entire current  $I$  pass through  $R_f$  as op-amp input current is zero.

Equating equation (1) & (2) we get

$$\frac{V_{in}}{R_1} = \frac{-V_O}{R_f}$$

$$\therefore A_{VF} = \frac{V_O}{V_{in}} = -\frac{R_f}{R_1} \quad (\text{Gain with feedback})$$

# Ideal Non-inverting Amplifier :- (3)



## Derivation of closed loop gain:-

→ The node B is at potential  $V_{in}$ , hence the potential of point A is same as B which is " $V_{in}$ ", from the concept of virtual short.

$$V_A = V_B = V_{in}$$



From the output side we can (4)

write

$$I = \frac{V_o - V_A}{R_f}$$

$$\therefore I = \frac{V_o - V_{in}}{R_f} \longrightarrow (1)$$

At the inverting terminal

$$I = \frac{V_A - 0}{R_1} \quad (\because V_o = 0)$$

$$I = \frac{V_A}{R_1} \quad (\because V_A = V_{in})$$

$$I = \frac{V_{in}}{R_1} \longrightarrow (2)$$

Entire current passes through  $R_1$  as input current of op-amp is zero

Equating eq (1) & (2) we get

(5)

$$\therefore \frac{V_o - V_{in}}{R_f} = \frac{V_{in}}{R_1}$$

$$\frac{V_o}{R_f} = \frac{V_{in}}{R_f} + \frac{V_{in}}{R_1}$$

$$\frac{V_o}{R_f} = V_{in} \left[ \frac{(R_1 + R_f)}{R_1 R_f} \right]$$

$$\frac{V_o}{V_{in}} = \frac{R_1 + R_f}{R_1}$$

$$\therefore A_{VF} = \frac{V_o}{V_{in}} = 1 + \frac{R_f}{R_1}$$